

1 Polarimetric Full-Wavefield Coherent Lidar

2 DONGYU DU,^{1,2,*} ANDREW XIE,^{1,2} PARSA MIRDEHGHAN,^{1,2}
3 BRANDON BUSCAINO,³ SEUNG-HWAN BAEK,⁴ KIRIAKOS N.
4 KUTULAKOS,^{1,2} DAVID B. LINDELL,^{1,2}

5 ¹ *Computer Science Department, University of Toronto, Toronto, Ontario, Canada*

6 ² *Vector Institute, Toronto, Ontario, Canada*

7 ³ *Ciena Corporation, San Jose, California, US*

8 ⁴ *Department of Computer Science and Engineering, Pohang University of Science and Technology*
9 *(POSTECH), Pohang, Gyeongsangbuk-do, Republic of Korea*

10 **dongyu@cs.toronto.edu*

11 **Abstract:** Coherent lidar recovers a scene’s geometry and motion by measuring the time-of-
12 flight and Doppler shift of backscattered light. However, conventional coherent lidar systems
13 do not capture the full optical wavefield—comprising time-varying amplitude, phase, and
14 polarization—and therefore forfeit access to an additional, rich source of scene information:
15 surface material properties encoded in the polarization state of backscattered light and in the
16 statistical properties of coherent polarization speckle. Prior approaches to full-wavefield sensing
17 require complex benchtop optics for ultrafast modulation and detection, or do not address the
18 modeling and recovery of the surface polarimetric response, which is scrambled by inter-reflections
19 within the lidar’s internal optics. Here, we introduce a polarimetric full-wavefield coherent
20 lidar system that recovers a surface point’s Jones matrix—encoding material and geometric
21 surface properties as observed through the system’s illumination and collection aperture—along
22 with depth and velocity, all from a single measurement. We achieve this by (1) repurposing
23 an off-the-shelf coherent optical modem—conventionally used for telecommunications—for
24 ultrafast full-wavefield modulation and detection, and (2) developing a novel, polarization-aware
25 image formation model and inverse method that disentangles the system’s optical response
26 from that of a target surface. Overall, the approach delivers millimeter-accurate depth, robust
27 velocimetry, and polarimetric surface properties, all with microsecond-scale exposures, eye-safe
28 optical power, and insensitivity to ambient light.

29 1. Introduction

30 Coherent lidar systems illuminate a scene with a modulated light source and interfere the
31 backscattered signal with a local oscillator to recover range and velocity. Compared to incoherent
32 lidar, coherent detection achieves shot-noise-limited sensitivity through coherent mixing gain [1]
33 and directly measures velocity via Doppler frequency shifts [2], making it widely used in
34 atmospheric sensing [3,4], robotic navigation [5], and long-range remote sensing [6].

35 In principle, coherent detection provides access to the full optical wavefield—amplitude, phase,
36 and polarization—which encodes not only range and velocity but also material-related properties
37 and micro-scale surface structure through the scene’s coherent polarimetric response (i.e., the
38 Jones matrix [7]) and the statistics of polarization speckle [8]. This full-wavefield information
39 could extend lidar beyond sensing geometry and motion to enable physical scene understanding,
40 which is important in autonomous driving [9–11], robotic manipulation [12–16] and maritime
41 sensing [17–20]. In practice, however, existing lidar systems do not fully exploit the full optical
42 wavefield due to challenges related to measurement acquisition and computational modeling.

43 The first challenge is acquiring full-wavefield measurements. Existing coherent lidar archi-
44 tectures either acquire only a subset of the wavefield or require complex custom optical setups.
45 Standard frequency-modulated continuous-wave (FMCW) systems [21,22] measure amplitude
46 and phase only in a single polarization channel; polarimetric FMCW variants [23] extend this to

47 multiple polarization states but demand additional beam synchronization optics and specialized
48 optical hardware. Random-modulated continuous-wave (RMCW) systems [24–26] typically rely
49 on either amplitude or phase modulation, but achieving millimeter-scale range resolution requires
50 ultrafast modulation at rates exceeding tens of gigahertz, which remains difficult to realize in
51 practice. Thus, neither architecture achieves a practical measurement front end for acquiring the
52 polarimetric full-wavefield response of the scene.

53 The second challenge is how to computationally model full-wavefield lidar measurements.
54 Even when coherent polarization information is available, existing systems lack the computational
55 framework to exploit it in the context of lidar. For example, coherent polarimetric sensing has
56 been demonstrated in adjacent modalities such as optical coherence tomography [27–29], but
57 extending these ideas to lidar requires sophisticated inverse modeling that accounts for Doppler
58 shifts, phase noise, speckle statistics, polarization-dependent scattering, and inter-reflections
59 within the system’s own optics.

60 Recently, Mirdehghan et al. [30] addressed the measurement acquisition challenge by re-
61 purposing a coherent optical modem—a device commonly used for high-bandwidth optical
62 communications [31]—for lidar sensing. Optical modems provide a nearly off-the-shelf platform
63 for ultrafast (>70 GHz) random amplitude and phase modulation in multiple polarization channels,
64 along with coherent detection of the full optical wavefield. Adapting one for lidar requires
65 only modest hardware changes: routing the output and input ports through a circulator and
66 coupling to free space through a collimator. Yet, despite these full-wavefield sensing capabilities,
67 the approach of Mirdehghan et al. does not recover the polarimetric response of a measured
68 surface. A key challenge is that the transmitted and received waveforms interact with optical
69 surfaces inside the lidar, coupling the system’s own polarization response with that of the scene.
70 Disentangling the two requires both a polarization-aware image formation model that accounts
71 for these internal interactions and a calibration procedure to characterize the system’s Jones
72 matrices.

73 Here, we repurpose a compact, off-the-shelf coherent optical modem to realize a full-wavefield
74 coherent lidar system with polarimetric sensing as shown in Fig. 1(a). The modem’s modulation
75 scheme is most closely related to RMCW lidar, but with the distinction that modulation is
76 applied jointly to both amplitude and phase across two orthogonal polarization states, enabling
77 full-wavefield sensing. Central to our approach is a polarization-aware image formation model
78 and inverse algorithm that separates the scene-dependent effective backscattering Jones response
79 from polarization distortions introduced by internal optical interfaces. We implement the
80 approach through a system calibration procedure and per-pixel offline optimization. We also
81 show that we can measure polarimetric speckle—whose properties correlate with micro-scale
82 surface roughness—and that we can infer material-related properties from the spatial degree
83 of polarization of the reflected field. Finally, we demonstrate using the polarimetric response
84 to distinguish between materials observed through a scattering layer. Overall, our approach
85 yields millimeter-accurate depth, robust velocimetry, and polarimetric surface characterization,
86 as shown in Fig. 1, all with microsecond-scale exposures, eye-safe optical power, and insensitivity
87 to ambient illumination.

88 2. Polarimetric Coherent Measurement Model

89 We present a model of the coherent modulation and detection process implemented by the optical
90 modem, and we analyze the effects introduced by partial polarization of backscattered light. The
91 coherent lidar system modulates both the amplitude and phase of light in two orthogonal linear
92 polarization channels. As illustrated in Fig. 1(a), this modulated wavefront is transmitted through
93 a fiber collimator, scanner, and circulator to illuminate the scene. After interacting with a surface,
94 the backscattered field is coupled back into the fiber, routed by the circulator to the modem and
95 recovered via homodyne detection in each channel. The returned signal consists of a component

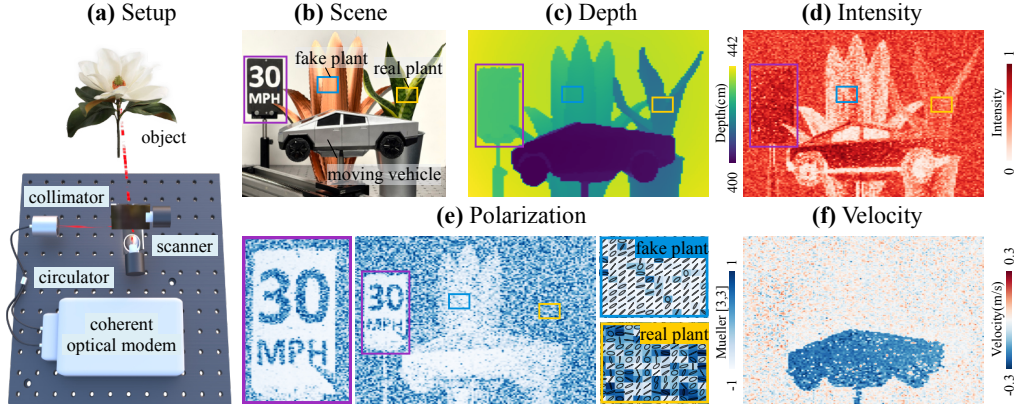


Fig. 1. **Simultaneous ranging, velocimetry, and polarimetric sensing with polarimetric full-wavefield coherent lidar.** (a) Experimental setup based on an off-the-shelf coherent optical modem, fiber collimator, circulator, and scanning mirrors for dual-polarization coherent measurement. (b) Experimental scene containing a moving model vehicle, a speed-limit sign, an artificial plant, and a real plant, mimicking a simplified road-side environment with both static and dynamic objects. (c) Reconstructed depth map, which localizes the spatial layout of the vehicle, sign, and surrounding vegetation. (d) Reconstructed intensity map, where the sign lettering is not resolved because the sign has nearly uniform albedo at the modem wavelength of 1550 nm. (e) Reconstructed polarization map, visualized by the Mueller M_{33} component and local polarization ellipses. The polarization map reveals material-sensitive details that are not resolved by depth or intensity, including the sign text and the difference between artificial and real vegetation. (f) Reconstructed Doppler velocity map, which identifies the moving vehicle and separates it from the static background.

96 that can be described by a Jones matrix, as well as a randomly polarized component that couples
 97 into the coherent detection process and gives rise to speckle. We provide an overview of this
 98 model in Fig. 2.

99 2.1. Modulation and Detection

100 **Modulation.** A narrow-linewidth continuous-wave laser provides the optical carrier at fre-
 101 quency f_0 . The coherent optical modem generates two known complex baseband waveforms
 102 in the orthogonal X and Y polarization channels. Specifically, the carrier is split into two
 103 polarization branches using a polarization beam splitter, and each branch is independently
 104 modulated by the modem-controlled dual-drive Mach-Zehnder modulators. The electrical
 105 symbol streams applied to the modulators control both the instantaneous amplitude and phase of
 106 each optical field component, producing the complex baseband signals $s_X(t) = A_X(t)e^{j\phi_X(t)}$ and
 107 $s_Y(t) = A_Y(t)e^{j\phi_Y(t)}$. Here, $A_X(t)$ and $A_Y(t)$ denote the time-varying field-envelope amplitudes,
 108 and $\phi_X(t)$ and $\phi_Y(t)$ denote the corresponding phase modulation waveforms in the two linear
 109 polarization channels. The two components are subsequently recombined to form $\mathbf{E}_{\text{tx}}(t)$, the
 110 transmitted dual-polarization optical field, which can be expressed as:

$$\mathbf{E}_{\text{tx}}(t) = \begin{bmatrix} A_X(t)e^{j\phi_X(t)} \\ A_Y(t)e^{j\phi_Y(t)} \end{bmatrix} e^{j2\pi f_0 t} = \mathbf{s}(t) e^{j2\pi f_0 t}, \quad (1)$$

111 where $\mathbf{s}(t) = [s_X(t), s_Y(t)]^T$ is the known dual-polarization complex baseband waveform
 112 generated by the optical modem and used as the probing signal in our forward model. Because

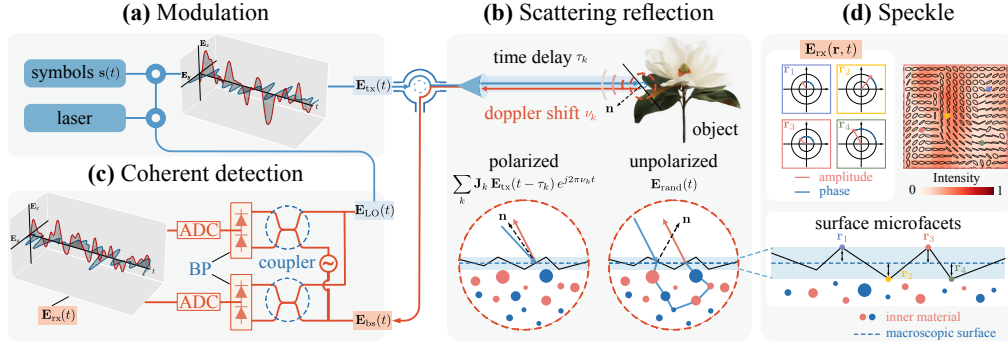


Fig. 2. **Polarimetric coherent measurement model.** (a) The baseband waveform $s(t)$ modulates the laser to synthesize a dual-polarization optical field $\mathbf{E}_{\text{tx}}(t)$. (b) The received field is a superposition of time-delayed (τ_k) and Doppler-shifted (ν_k) polarized echoes, together with a randomly polarized component $\mathbf{E}_{\text{rand}}(t)$ arising from multiple scattering. (c) The returned signal $\mathbf{E}_{\text{bs}}(t)$ interferes with a local oscillator $\mathbf{E}_{\text{LO}}(t)$ in a balanced receiver; bandpass filtering and analog-to-digital (ADC) sampling yield complex baseband measurements $\mathbf{E}_{\text{rx}}(t)$ in two polarization channels. (d) The measured field $\mathbf{E}_{\text{rx}}(\mathbf{r}, t)$ varies across spatial locations \mathbf{r} due to coherent interference among surface microfacets, resulting in intensity and polarization speckle.

113 both polarization channels are independently modulated in amplitude and phase, $s(t)$ provides a
 114 time-varying set of input polarization states for estimating the polarimetric scene response.

115 **Free-space coupling and surface backscattering.** After modulation, $\mathbf{E}_{\text{tx}}(t)$ is transmitted
 116 through a single-mode fiber, an amplifier, and an optical circulator (see Fig. 2) before being
 117 collimated into free space. Then, the backscattered light from the scene is subsequently recollimated
 118 by the same collimator and routed through the circulator to the receiver port of the optical
 119 modem. We model the received waveform by accounting for (1) round-trip propagation delays
 120 from internal and scene reflections, (2) polarization changes induced by the fiber, internal optics,
 121 and scene surface, (3) Doppler-induced carrier-frequency shifts, and (4) a random, unpolarized
 122 field generated by surface scattering. The backscattered field $\mathbf{E}_{\text{bs}}(t)$ is thus written as:

$$\mathbf{E}_{\text{bs}}(t) = \sum_k \mathbf{J}_k \mathbf{E}_{\text{tx}}(t - \tau_k) e^{j2\pi\nu_k t} + \mathbf{E}_{\text{rand}}(t). \quad (2)$$

123 Here, \mathbf{J}_k is the Jones matrix corresponding to the k -th surface reflection, which models a
 124 deterministic polarization transformation; τ_k and ν_k represent the corresponding time delay and
 125 Doppler frequency shift; and $\mathbf{E}_{\text{rand}}(t)$ is a random, unpolarized component of the optical field
 126 that cannot be represented by a single Jones matrix transformation.

127 **Coherent detection.** The backscattered field $\mathbf{E}_{\text{bs}}(t)$ is routed by the circulator to the coherent
 128 receiver of the optical modem. Inside the modem receiver, the returned signal is combined
 129 with a strong local-oscillator (LO) field $\mathbf{E}_{\text{LO}}(t)$ using an optical hybrid. For each polarization
 130 channel, the hybrid produces in-phase and quadrature interference outputs, which are measured
 131 by balanced photodetectors and digitized to form the complex baseband samples. Under the
 132 strong-LO approximation [32], the differential photocurrent is proportional to the interference
 133 term between the LO and the received optical field, yielding

$$\mathbf{E}_{\text{rx}}(t) = \kappa \mathbf{E}_{\text{LO}}^\dagger(t) \left(\sum_k \mathbf{J}_k \mathbf{E}_{\text{tx}}(t - \tau_k) e^{j2\pi\nu_k t} + \mathbf{E}_{\text{rand}}(t) \right) + \boldsymbol{\eta}(t), \quad (3)$$

134 where $(\cdot)^\dagger$ denotes the Hermitian transpose, κ is the heterodyne mixing constant [33, 34], and
 135 $\boldsymbol{\eta}(t)$ denotes additive electronic and shot noise.

136 2.2. Coherency and Polarization Speckle

137 Up to this point, we have modeled the received electric field, \mathbf{E}_{rx} , as the sum of a coherent
 138 component described by interaction with the Jones matrices \mathbf{J}_k and random components captured
 139 by \mathbf{E}_{rand} and $\boldsymbol{\eta}(t)$. We now describe how to use the coherency matrix [7] to model the fraction of
 140 randomly polarized light in \mathbf{E}_{rx} , and we explain how the coherent interactions of light give rise to
 141 polarization speckle.

142 **Coherency matrix.** To analyze the polarization characteristics of \mathbf{E}_{rx} , we analyze its second-
 143 order coherence through the coherency matrix: $\mathbf{C}_{\text{out}} = \mathbb{E}[\mathbf{E}_{\text{rx}}(t)\mathbf{E}_{\text{rx}}^\dagger(t)]$, where $\mathbb{E}[\cdot]$ denotes time
 144 averaging. \mathbf{C}_{out} is a 2×2 Hermitian positive-semidefinite matrix that describes the amplitude
 145 and phase correlations between orthogonal polarization components. A statistical representation
 146 of polarized light can then be obtained by the eigen decomposition of \mathbf{C}_{out} :

$$\mathbf{C}_{\text{out}} = \lambda_1 \mathbf{v}_1 \mathbf{v}_1^\dagger + \lambda_2 \mathbf{v}_2 \mathbf{v}_2^\dagger, \quad \lambda_1 \geq \lambda_2 \geq 0, \quad (4)$$

147 where $\lambda_{1/2}$ and $\mathbf{v}_{1/2}$ are the eigenvalues and eigenvectors of \mathbf{C}_{out} . In Section 3, we introduce a
 148 statistical optimization model for estimating the fully polarized Jones term \mathbf{J}_k and the incoherent
 149 power which is proportional to the unpolarized component λ_2 .

150 **Polarization speckle.** In practice, coherent lidar measurements exhibit speckle because micron-
 151 scale variations in surface geometry create differing optical path lengths in the backscattered
 152 light [8].

153 We can describe this phenomenon by modeling the received field at detector position \mathbf{r} and
 154 time t , which we denote $\mathbf{E}_{\text{rx}}(\mathbf{r}, t)$. This field is the coherent superposition of contributions from
 155 many micro-facets over the illuminated surface area \mathcal{A} :

$$\mathbf{E}_{\text{rx}}(\mathbf{r}, t) = \int_{\mathcal{A}} \mathbf{J}(\mathbf{r}') \mathbf{E}_{\text{tx}}(\mathbf{r}', t) e^{jk_0(|\mathbf{r}-\mathbf{r}'|+h(\mathbf{r}'))} d\mathbf{r}' + \mathbf{E}_{\text{rand}}(\mathbf{r}, t). \quad (5)$$

156 Here, $\mathbf{J}(\mathbf{r}')$ denotes the local Jones matrix at surface point \mathbf{r}' , $h(\mathbf{r}')$ represents the micron-scale
 157 surface height profile, and k_0 is the optical wavenumber. The phase term encodes path-length
 158 variations caused by changes in surface height (Fig. 2(d)); increasing roughness (surface height
 159 variance) causes phase decorrelation between micro-facets, which in turn reduces the spatial
 160 degree of polarization and yields increasingly randomized polarization speckle.

161 3. Scene Reconstruction

162 We first introduce a statistical optimization model that jointly estimates Jones responses and
 163 incoherent power from the received field $\mathbf{E}_{\text{rx}}(t)$. Then, we show how to derive the depth and
 164 velocity based on the estimated Jones responses. Finally, we decompose the recovered Jones
 165 matrix into separate components that model the surface reflection, internal reflections and
 166 dispersion effects within the lidar system.

167 **Statistical optimization model.** Eq. (3) shows that the received coherent field $\mathbf{E}_{\text{rx}}(t)$ consists
 168 of a deterministic component induced by the modulated RMCW transmitted waveform, a random
 169 scattering component and additive receiver noise. Therefore, the measurement follows a complex
 170 Gaussian process: $\mathbf{E}_{\text{rx}}(t) \sim \mathcal{CN}(\boldsymbol{\mu}(t), \boldsymbol{\Sigma}(t))$, as shown in Fig. 3(a). The mean value $\boldsymbol{\mu}(t)$ and
 171 covariance $\boldsymbol{\Sigma}(t)$ can be expressed as follows:

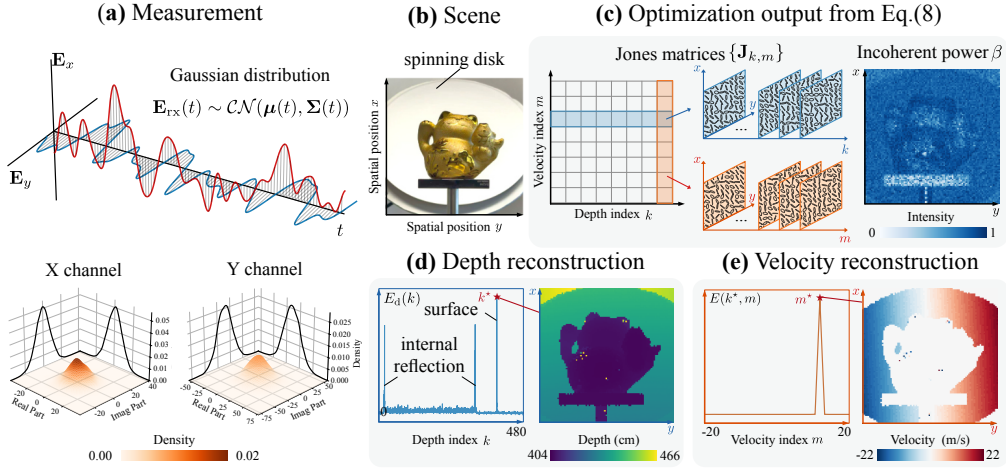


Fig. 3. **Scene reconstruction overview.** (a) The two-channel measurement $\mathbf{E}_{\text{rx}}(t)$ is modeled as a complex Gaussian random process; the bottom plots show the empirical complex-Gaussian densities of the X and Y polarization channels. (b) Example scene capture (a spinning disk and a figurine) scanned over spatial coordinates (x, y) . (c) We estimate per-pixel, depth–Doppler-indexed Jones matrices $\{\mathbf{J}_{k,m}\}$ together with an incoherent power β by minimizing the negative log-likelihood in Eq. (8). (d) We compute the measurement energy using the Frobenius norm of the Jones matrix, $E(k, m) = \|\mathbf{J}_{k,m}\|_F^2$, integrate over Doppler to obtain the depth profile $E_d(k)$, and estimate depth as $k^* = \arg \max_k E_d(k)$. (e) We obtain velocity by identifying the dominant Doppler index $m^* = \arg \max_m E(k^*, m)$.

$$\boldsymbol{\mu}(t) = \kappa \mathbf{E}_{\text{LO}}^\dagger(t) \sum_k \mathbf{J}_k \mathbf{E}_{\text{tx}}(t - \tau_k) e^{j2\pi\nu_k t}, \quad \boldsymbol{\Sigma}(t) = (\sigma_{\text{het}}^2(t) + \beta(t)) \mathbf{I}_2. \quad (6)$$

172 The value $\sigma_{\text{het}}(t)$ denotes the heterodyne noise variance [31], $\beta(t) = |\kappa|^2 \|\mathbf{E}_{\text{LO}}\|^2 \lambda_2(t)$
 173 corresponds to the incoherent power of the measured field, and $\lambda_2(t)$ is the second eigenvalue
 174 of the local coherency matrix in Eq. (4). A derivation of Eq. (6) is provided in Supplement 1,
 175 Section 1.A.

176 Directly optimizing the continuous Doppler shifts $\{\nu_k\}$ jointly with the Jones kernels $\{\mathbf{J}_k\}$
 177 and the incoherent power term $\beta(t)$ leads to a highly non-convex problem. To obtain a stable
 178 and efficient solver, we discretize the Doppler axis onto a fixed grid $\{\nu_m\}_{m=1}^M$ and estimate
 179 a depth–velocity Jones response volume $\{\mathbf{J}_{k,m}\}$, as shown in Fig. 3(c). Accounting for this
 180 discretization, the mean value in Eq. (6) becomes

$$\boldsymbol{\mu}(t) = \kappa \mathbf{E}_{\text{LO}}^\dagger(t) \sum_k \sum_{m=1}^M \mathbf{J}_{k,m} \mathbf{E}_{\text{tx}}(t - \tau_k) e^{j2\pi\nu_m t}. \quad (7)$$

181 Assuming independent samples in time, we estimate the depth–velocity Jones response volume,
 182 and the incoherent power term by minimizing the negative log-likelihood of the corresponding
 183 Gaussian distribution. Additionally, we introduce a sparsity-promoting regularization term to
 184 stabilize the ill-posed inversion and suppress spurious depth and velocity responses. The resulting
 185 objective function is given as

$$\min_{\{\mathbf{J}_{k,m}\}, \beta} \sum_t \left(2 \log q + \frac{\|\mathbf{E}_{\text{rx}}(t) - \boldsymbol{\mu}(t)\|_2^2}{q} \right) + \lambda_{\text{reg}} \sum_{k,m} \|\mathbf{J}_{k,m}\|_F, \quad (8)$$

186 where $\|\cdot\|_F$ denotes the Frobenius norm of the Jones matrix, λ_{reg} controls the strength of
 187 the regularization, and q is the variance term in the diagonal covariance $\Sigma(t)$. **A stability**
 188 **analysis of this scalar-covariance approximation under anisotropic depolarized covariance and**
 189 **polarization-dependent noise is provided in Supplement 1, Section 1.D.** Over each acquisition
 190 window, we assume approximately stationary noise statistics and therefore adopt a time-invariant
 191 variance $q = \sigma_{\text{het}}^2 + \beta$. Here σ_{het}^2 is obtained via a noise-floor estimation from a measurement
 192 window without target returns (see details in Supplement 1, Section 1.B), and $\beta \geq 0$ captures the
 193 additional variance induced by the incoherent power.

194 **Depth and velocity estimation.** To recover depth and Doppler, we convert the estimated depth–
 195 velocity Jones response volume $\{\mathbf{J}_{k,m}\}$ to an energy map by computing the Frobenius norm of
 196 the Jones matrices: $E(k, m) = \|\mathbf{J}_{k,m}\|_F^2$. We then integrate over the Doppler bins to form a 1D
 197 depth profile $E_d(k)$ and pick the target delay index as the strongest return, $k^\star = \arg \max_k E_d(k)$.
 198 Given the selected depth k^\star , we estimate the Doppler index from the corresponding velocity
 199 profile by $m^\star = \arg \max_m E(k^\star, m)$, yielding the final depth and velocity estimates τ_{k^\star} and v_{m^\star} .
 200 Repeating this procedure over all spatially scanned pixels produces the depth map and velocity
 201 map shown in Fig. 3(d–e).

202 **Decomposing the surface-reflection Jones matrix.** The depth–velocity index (k^\star, m^\star)
 203 identifies the Jones matrix $\mathbf{J}^\star \triangleq \mathbf{J}_{k^\star, m^\star}$, which represents the combined round-trip polarization
 204 response of the system, including the transmit-path optics \mathbf{J}_{tx} , the surface reflection \mathbf{J}_{surf} , and the
 205 receive-path optics \mathbf{J}_{rx} . To recover the surface polarization, we separately model the response of
 206 the transmitted and received optical paths and computationally remove their effects. Once \mathbf{J}_{tx}
 207 and \mathbf{J}_{rx} are calibrated (described in Section 4), the decoupled surface Jones matrix is obtained as:

$$\mathbf{J}_{\text{surf}} = \gamma^{-1} \mathbf{J}_{\text{rx}}^{-1} \mathbf{J}^\star \mathbf{J}_{\text{tx}}^{-1}. \quad (9)$$

208 where $\gamma \in \mathbb{R}_+$ denotes a global amplitude gain.

209 **\mathbf{J}_{surf} is an effective backscattering Jones response observed through the system’s illumination**
 210 **and collection aperture.** We analyze its **effective polarimetric descriptors**—including diattenuation
 211 and retardance—using the standard polar decomposition [35]. These parameters provide
 212 physically interpretable cues for material discrimination and surface characterization. Beyond
 213 per-pixel properties, \mathbf{J}_{surf} also exhibits rich spatial variations driven by micro-geometry and
 214 coherent interference, which manifest as polarization speckle patterns across the scanned field
 215 of view. We analyze the spatial structure of polarization speckle using the spatial degree of
 216 polarization (SDoP) [8, 36], which we show is useful to compare materials and surface roughness.
 217 We provide precise definitions of these quantities (diattenuation, retardance, and SDoP) in
 218 Supplement 1, Section 1.C. **We further analyze the stability of the joint depth–velocity–Jones**
 219 **reconstruction under random initialization in Supplement 1, Section 1.F.**

220 4. Polarimetric Calibration

221 As illustrated in Fig. 4(a), the polarization state measured by the coherent receiver is shaped not
 222 only by the scene, but also by polarization distortions accumulated along the transmitted and
 223 received optical paths, including the fiber link, erbium-doped fiber amplifier (EDFA), circulator,
 224 collimator and scanner. We abstract these path-dependent distortions as two Jones operators, \mathbf{J}_{tx}
 225 and \mathbf{J}_{rx} , which compactly represent the cascaded polarization transformations of the transmitted
 226 and received paths, respectively. Accordingly, we design a calibration setup and an end-to-end
 227 optimization procedure to jointly estimate \mathbf{J}_{tx} and \mathbf{J}_{rx} from a set of captured measurements.

228 First, we insert a rotating linear polarizer in the common transmitted/received path to generate
 229 controllable polarization states at the target and thereby obtain multiple linearly independent

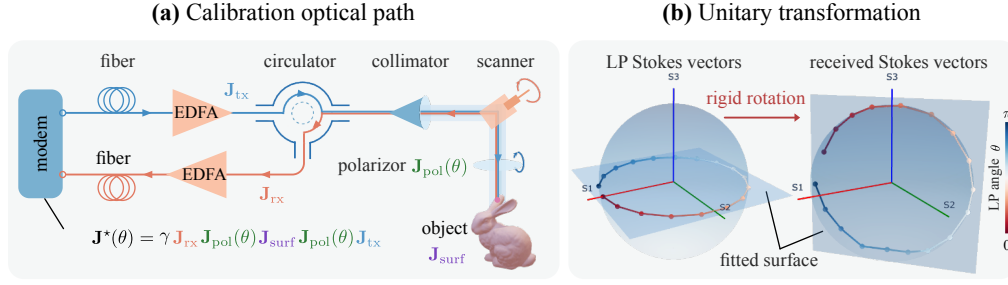


Fig. 4. **Polarimetric calibration overview.** (a) A rotating linear polarizer with Jones operator $\mathbf{J}_{\text{pol}}(\theta)$ is inserted in the common transmitted/received path to generate known polarization states. The measured response is modeled as $\mathbf{J}^*(\theta) = \gamma \mathbf{J}_{\text{rx}} \mathbf{J}_{\text{pol}}(\theta) \mathbf{J}_{\text{surf}} \mathbf{J}_{\text{pol}}^\dagger(\theta) \mathbf{J}_{\text{tx}}$, where \mathbf{J}_{tx} and \mathbf{J}_{rx} represent the polarization distortions introduced by the transmitted and received optical paths of the optical train (fiber, EDFA, circulator, collimator and scanner), respectively. \mathbf{J}_{surf} is the Jones matrix of the calibration object. (b) As θ varies, the Stokes vectors of the linear polarization (LP) states trace a circle on the Poincaré sphere; after propagation through the system, the Stokes vectors remain on a rotated circle, consistent with a unitary polarization transformation.

230 measurements during calibration. Note that the polarizer is used only during calibration and
 231 removed for all subsequent imaging experiments. For a polarizer angle θ , the estimated combined
 232 surface response $\mathbf{J}(\theta)$ is

$$\mathbf{J}(\theta) = \gamma \mathbf{J}_{\text{rx}} \mathbf{J}_{\text{pol}}(\theta) \mathbf{J}_{\text{surf}} \mathbf{J}_{\text{pol}}^\dagger(\theta) \mathbf{J}_{\text{tx}}. \quad (10)$$

233 We define $\mathbf{J}_{\text{pol}}(\theta)$ to be the Jones operator of a linear polarizer at angle θ [37]:

$$\mathbf{J}_{\text{pol}}(\theta) = \mathbf{R}(\theta) \mathbf{P}_0 \mathbf{R}^\top(\theta), \quad \mathbf{R}(\theta) = \begin{bmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{bmatrix}, \quad \mathbf{P}_0 = \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix}. \quad (11)$$

234 The matrix $\mathbf{R}(\theta)$ denotes a rotation operator that aligns the linear polarizer's transmission axis
 235 with the measurement coordinate frame, and \mathbf{P}_0 represents an ideal horizontal linear polarizer.
 236 For the calibration target we assume a homogeneous, non-depolarizing planar surface whose
 237 Jones matrix can be diagonalized in its principal polarization basis, so we parameterize it as the
 238 rotated diagonal form:

$$\mathbf{J}_{\text{surf}} = \mathbf{R}(\alpha)^\top \begin{bmatrix} r_1 e^{j\phi_1} & 0 \\ 0 & r_2 e^{j\phi_2} \end{bmatrix} \mathbf{R}(\alpha), \quad (12)$$

239 with amplitude–phase coefficients (r_1, ϕ_1) and (r_2, ϕ_2) in the principal axes and an in-plane
 240 orientation α .

241 **Unitary approximation.** Polarization evolution in single-mode fibers and reciprocal passive
 242 optics can be approximated by a unitary Jones operator [38]. Geometrically, a unitary Jones
 243 transformation corresponds to a rigid rotation on the Poincaré sphere; thus, the locus traced by a
 244 rotating linear polarizer forms an equatorial circle that remains invariant under the transformation.
 245 Our experimental observations in Fig. 4(b) further confirm this assumption (see Supplement 1,
 246 Section 2.A for a derivation and details). We therefore parameterize \mathbf{J}_{tx} and \mathbf{J}_{rx} as a rotation

247 followed by a linear retarder:

$$\mathbf{U}(\psi, \delta, \vartheta) = \mathbf{R}(\psi) \mathbf{L}_{\delta, \vartheta}, \quad \mathbf{L}_{\delta, \vartheta} = \begin{bmatrix} \cos \frac{\delta}{2} - j \sin \frac{\delta}{2} \cos 2\vartheta & -j \sin \frac{\delta}{2} \sin 2\vartheta \\ -j \sin \frac{\delta}{2} \sin 2\vartheta & \cos \frac{\delta}{2} + j \sin \frac{\delta}{2} \cos 2\vartheta \end{bmatrix}, \quad (13)$$

248 where ψ , δ , and ϑ denote the global rotation, retardance, and fast-axis orientation, respectively.
 249 Assigning $\mathbf{J}_{\text{rx}} = \mathbf{U}(\psi_t, \delta_t, \vartheta_t)$ and $\mathbf{J}_{\text{tx}} = \mathbf{U}(\psi_r, \delta_r, \vartheta_r)$ ensures $\mathbf{U}^\dagger \mathbf{U} = \mathbf{I}$.

250 **Optimization.** Given a set of measured Jones responses $\{\mathbf{J}^*(\theta_n)\}_{n=1}^N$ acquired at different linear
 251 polarizer angles $\{\theta_n\}$, the decomposition problem is formulated as a constrained optimization:

$$\min_{\Theta} \sum_{n=1}^N \|\mathcal{M}(\mathbf{J}(\theta_n; \Theta)) - \mathcal{M}(\mathbf{J}^*(\theta_n))\|_F^2, \quad (14)$$

252 where $\mathcal{M}(\cdot)$ denotes the Jones-to-Mueller mapping [35] and $\Theta = \{\gamma, \psi_{t/r}, \delta_{t/r}, \vartheta_{t/r}, r_{1/2}, \phi_{1/2}, \alpha\}$
 253 collects all unknowns.

254 We use the Jones-to-Mueller mapping to account for the global phase offsets present in our
 255 measured Jones matrices. In practice, each Jones response $\mathbf{J}^*(\theta_n)$ is acquired at a different
 256 time and is therefore multiplied by an unknown global phase factor arising from interferometric
 257 phase drift between the local oscillator and signal, laser frequency noise, and path-length
 258 fluctuations [1]. However, the polarization state depends only on the relative phase between the
 259 orthogonal polarization channels. Consequently, measurement-to-measurement variations in
 260 this global phase offset can cause our optimization (Eq. (14)) to fail if we penalize discrepancies
 261 directly in the Jones matrix domain. In contrast, the corresponding Mueller matrices $\mathcal{M}(\mathbf{J})$ are
 262 invariant to global phase, and therefore keep the relative amplitudes and relative phase between
 263 the two orthogonal polarization components unchanged. Defining the loss in the Mueller domain
 264 therefore resolves the phase ambiguity. A derivation and illustrative example are provided in
 265 Supplement 1, Section 2.B.

266 5. Results

267 5.1. Hardware Prototype

268 As shown in Fig. 5, the experimental lidar prototype employs a Ciena WaveLogic 5n optical
 269 modem operating at a wavelength of 1550 nm. The modem samples the optical field at 74 GHz,
 270 yielding an optical depth resolution of 2 mm. The transmitted sequence length is constrained
 271 by the internal modem memory to approximately 2^{16} symbols, corresponding to an exposure
 272 time of 0.89 μs . Waveform programming and data acquisition are performed via the QSFP-DD
 273 electrical interface. An Erbium-doped fiber amplifier (Thorlabs EDFA300S) is incorporated in
 274 the transmitted path to increase the optical power. During operation, the average transmitted
 275 power is 2.1 mW, well below the Class 1 limit at 1550 nm (~ 10 mW, continuous wave). A
 276 second amplifier (Thorlabs EDFA100S) is deployed in the received path to pre-amplify the
 277 backscattered signal to roughly 0.9 mW, ensuring optimal input power for the modem. A Thorlabs
 278 CFS18-1550-APC fiber collimator lens converts the fiber output into a low-divergence free-space
 279 beam, maintaining beam quality across the working distance. The outgoing beam is deflected by
 280 a Thorlabs GVS012 dual-axis scanning galvanometer, enabling precise angular control for raster
 281 scanning and three-dimensional scene acquisition.

282 5.2. Experimental Results

283 5.2.1. Evaluation of Polarization–Depth–Velocity Reconstruction

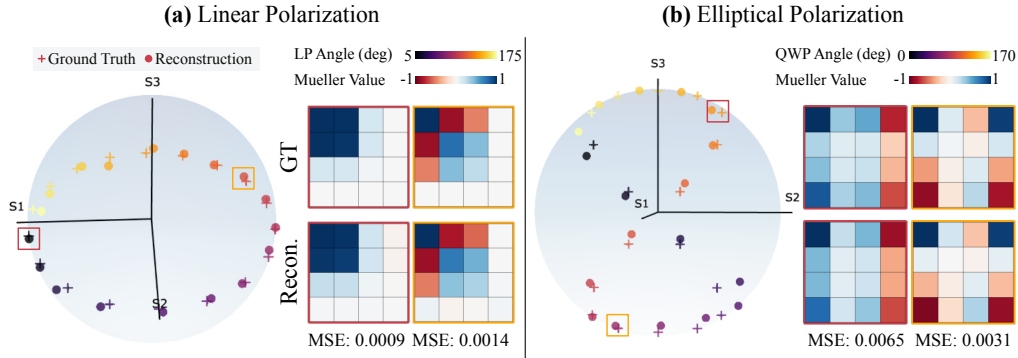


Fig. 6. **Polarization calibration results.** (a) We rotate the linear polarizer to generate a set of linearly polarized input states (Stokes vectors) and compare the ground truth (cross) with the reconstruction (dot) on the Poincaré sphere. Representative Mueller-matrix reconstructions for two of these states (red and yellow squares) are shown with their mean squared error (MSE). (b) We place a QWP after the linear polarizer and rotate the QWP to produce elliptical states; the recovered Stokes trajectory and two example Mueller matrices are shown in the same manner.

284 **Polarization calibration evaluation.** To evaluate polariza-
 285 tion recovery accuracy, we perform two controlled tests: (1)
 286 a linear polarizer (LP) rotated to 18 angles from 5° to 175°
 287 (excluding calibration angles), and (2) a cascaded LP(0°)
 288 plus quarter-wave plate (QWP) with the QWP rotated from
 289 0° to 170° in 18 steps. For each case, we convert the re-
 290 constructed Jones matrices to Stokes vectors and plot them
 291 on the Poincaré sphere (Fig. 6), comparing reconstructed
 292 states (dots) to analytical ground truth from ideal LP/QWP
 293 models (crosses). The recovered trajectories closely follow
 294 the theoretical loci for both linear and elliptical states. Fig. 6
 295 also reports representative reconstructed Mueller matrices,
 296 showing close agreement with ground truth across all el-
 297 ements ($MSE \sim 10^{-3}$). A comparison with and without
 298 polarization calibration is also provided in Supplement 1,
 299 Section 5.A. We also evaluate the calibration efficiency and
 300 repeatability in Supplement 1, Section 2.C.

301 In addition to standard polarization elements, we compare
 302 the reconstructed Mueller matrix of a 3D object against
 303 physics-based polarimetric simulations [39] under matched
 304 configurations, which shows good qualitative agreement
 305 (details are provided in Supplement 1, Section 4).

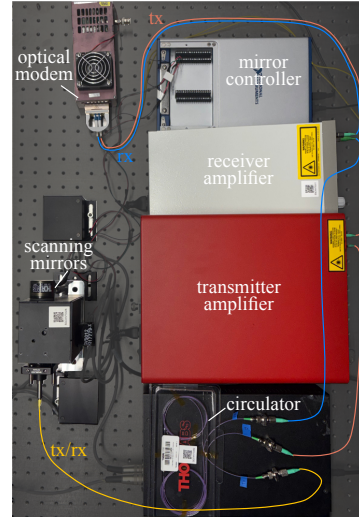


Fig. 5. **Hardware prototype.** The system consists of a coherent optical modem, transmitter and receiver EDFAs, a fiber circulator, scanning mirrors, and a controller.

306 **Depth and velocity reconstruction across algorithms.** We compare the proposed method
 307 with a matched-filter baseline and the FWL approach [30]. The matched filter fails to accurately
 308 recover depth for moving objects and cannot estimate radial velocity. Both FWL and our method
 309 jointly reconstruct depth and velocity; however, by explicitly accounting for incoherent power in
 310 a likelihood-based formulation, our method is more robust in low-SNR regions and achieves the
 311 lowest Mean Absolute Error (MAE) for both depth (4.37 mm) and radial velocity (0.86 m/s), as
 312 shown in Fig. 7. Additionally, we accelerate the forward model using FFT-based convolution,

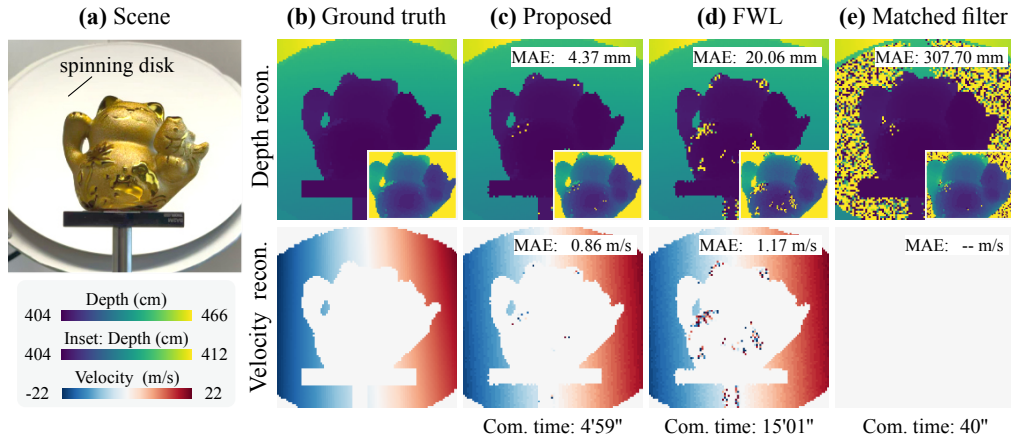


Fig. 7. **Comparison of depth and velocity reconstruction across algorithms.** (a) Experimental scene with a figurine and a spinning disk. (b) Ground-truth depth and radial velocity maps (depth from high-power acquisition; velocity from a high-speed camera, see details in Supplement 1, Section 5.B). (c–e) Reconstructed depth (top) and radial velocity (bottom) using the proposed method, FWL [30], and a matched-filter baseline. MAE and computation time are reported in each column. The matched-filter baseline cannot estimate radial velocity.

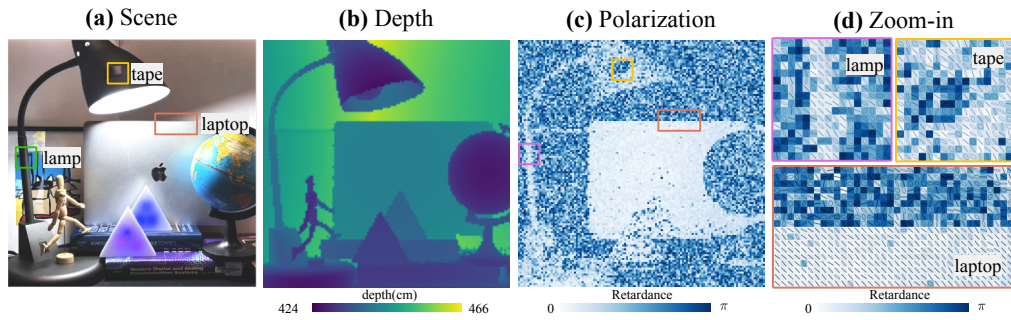


Fig. 8. **Reconstruction under strong ambient light.** (a) A tabletop scene is lit by multiple uncontrolled sources (desk lamp, ceiling light, triangular-shaped blue and purple LEDs), creating strong spatially-varying background light. (b) Despite the presence of ambient light, coherent detection enables robust depth recovery with clear object geometry. (c) The reconstructed polarimetric response remains accurate under ambient illumination. (d) The zoomed-in insets reveal metallic–dielectric differences in both retardance and the overlaid Jones ellipse features (orientation and ellipticity), comparing the lamp arm/head and laptop with the background paper and tape.

313 making our implementation approximately three times faster than FWL. All computation times
 314 are reported as wall-clock time measured on an Intel Xeon Gold 5120 CPU and an NVIDIA
 315 TITAN RTX GPU. Additional analyses are provided in Supplement 1, Section 5.B. **We further**
 316 **evaluate reconstruction performance over varying target depths, target velocities, and SNR levels**
 317 **in Supplement 1, Section 5.C.**

318 5.2.2. Reconstruction under Ambient Light

319 To demonstrate robustness under realistic, uncontrolled indoor lighting conditions, we capture
 320 scenes in a brightly lit indoor environment (about 10^4 lux), under simultaneous illumination from

321 desk lamps, ceiling lights, triangular-shaped blue and purple LEDs (Fig. 8(a)).

322 Despite the strong illumination, our coherent polarimetric pipeline still jointly reconstructs
323 accurate depth and polarization information. Fig. 8(b) shows that the recovered depth preserves
324 the scene geometry, while the reconstructed Jones retardance map in Fig. 8(c) remains well-
325 structured and retains material-discriminative signatures. The zoom-ins in Fig. 8(d) highlight
326 three representative regions: (1) the lamp arm against the picture-frame background, (2) the lamp
327 head with an attached tape patch, and (3) the laptop surface against a paper background. Across
328 these examples, metallic regions (lamp and laptop) appear spatially smooth, with locally correlated
329 retardance and consistent Jones-ellipse orientation and ellipticity, whereas the dielectric regions
330 (tape and background) exhibit noticeably higher spatial variability with patchier, speckle-like
331 fluctuations in both retardance and the recovered Jones-ellipse patterns.

332 5.2.3. Polarization Coherent Speckle Reconstruction Results

333 **Polarization speckle reconstruction for different materials.** As shown in Fig. 9(a–b), we
334 investigate the polarization-coherent speckle characteristics across different materials: a painted
335 metallic car body panel, a translucent wax block, a 3D-printed polymer figurine (PLA), an
336 artificial (plastic) plant, and a real plant leaf. For each target we scan a 2 mm by 2 mm patch and
337 reconstruct both the intensity speckle and the associated polarization field (Stokes vectors and
338 SDoP). Fig. 9(c) compares the reconstructed intensity speckle overlaid with polarization speckle
339 for all materials. While the intensity speckle patterns appear random and visually similar, their
340 polarization structures are markedly different.

341 For metallic surfaces, their large imaginary refractive index confines light within a shallow skin
342 depth and maintains a nearly constant phase relationship between the two orthogonal polarized
343 components, resulting in spatially coherent polarization states with the highest SDoP. Dielectric
344 3D-printed surfaces exhibit stronger polarization mixing and angle-dependent Fresnel effects,
345 producing lower SDoP and more rapidly varying polarization speckle. Translucent wax shows
346 intermediate behavior: subsurface scattering reduces SDoP relative to metal but still preserves
347 smoother polarization structures than the rough dielectric. Finally, the real plant leaf exhibits
348 strongly randomized polarization with low SDoP due to multiple volume-scattering layers,
349 whereas the artificial plant—made of smooth polymer—produces polarization speckle patterns
350 similar to the 3D-printed object. These observations demonstrate that polarization-coherent
351 speckle encodes rich, material-dependent scattering signatures that are not discernible from
352 intensity-only measurements.

353 **Effect of varying surface roughness on polarization speckle.** As described by the polarization
354 speckle model in Sec. 2.2, surface height variations modulate the optical path length and introduce
355 a spatially-varying phase term. As the variance of surface height increases, the path length
356 fluctuations decorrelate the field between neighboring micro-facets, reducing the local SDoP and
357 making the polarization speckle more random.

358 To experimentally verify this relationship, we measure polarization speckle on a set of
359 sandpaper samples with calibrated roughness levels (70–270 μm grain size). For each sample,
360 we reconstruct the polarization-resolved Jones matrix field over a 2 mm by 2 mm patch and
361 visualize the resulting polarization speckle, together with the corresponding SDoP. As shown
362 in Fig. 9(c), as the surface roughness increases (corresponding to a larger grain size), SDoP
363 decreases monotonically, indicating stronger depolarization and higher speckle randomness.

364 We further investigate these trends in Supplement 1, Section 3, where we analyze how
365 polarization speckle statistics and SDoP vary with the system aperture size and scanned field-of-
366 view. These results show that larger apertures and wider scan areas encompass more independent
367 scatterers, which strengthens speckle randomness and further reduces the measured SDoP.

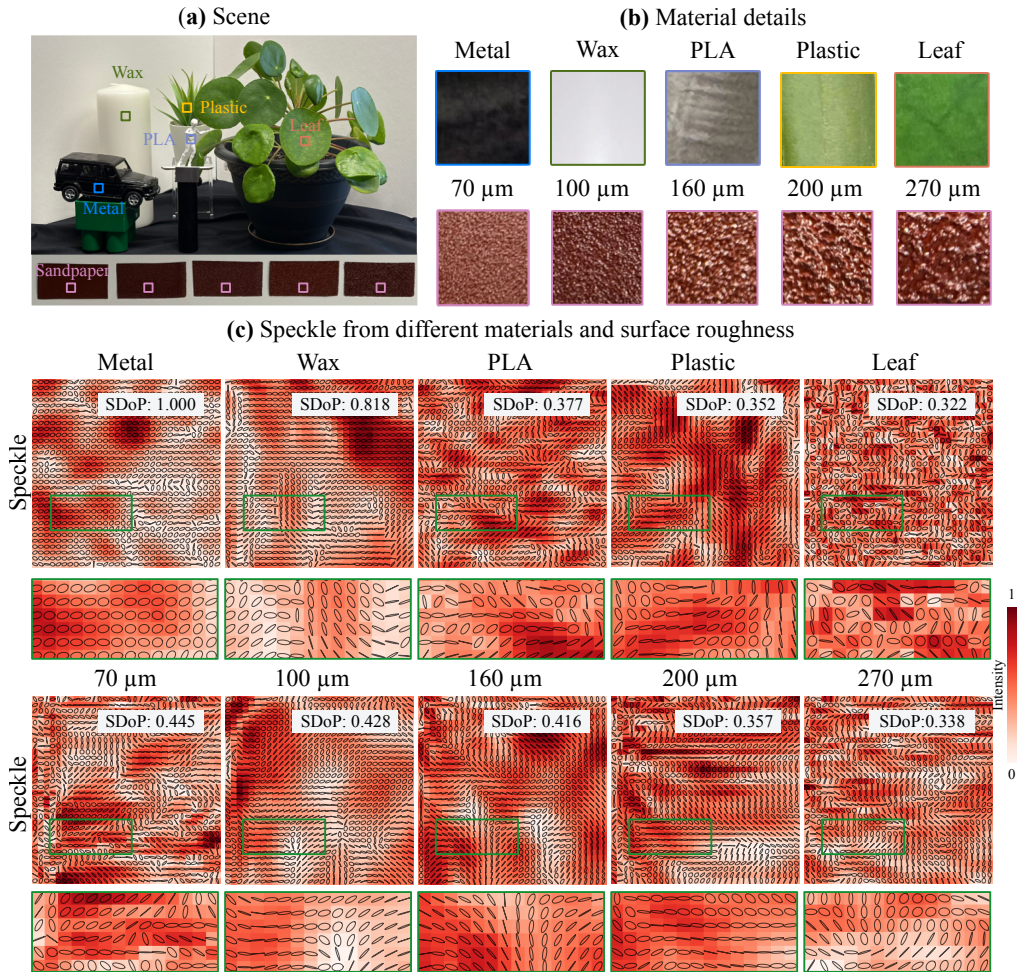


Fig. 9. **Polarization speckle from different materials and surface roughness.** (a–b) Test scene and surface details for five representative materials (metal, wax, PLA, plastic, and leaf) together with sandpaper samples of increasing grain size (70–270 μm). (c) For each material and sandpaper sample, we visualize the reconstructed intensity speckle overlaid with polarization ellipses over an approximately 2 mm by 2 mm scan, along with a zoomed-in patch. The SDoP decreases from metal to translucent wax and dielectric PLA/plastic due to increasing subsurface scattering that progressively randomizes polarization. The real leaf, dominated by multiple scattering within complex biological microstructures, exhibits highly random polarization states. For sandpaper, the SDoP decreases monotonically as the surface becomes rougher.

368 5.2.4. Polarimetric Imaging Through Scattering Media

369 Polarimetric information can provide material-sensitive contrast in imaging-through-scattering
 370 settings where intensity contrast is weak. To evaluate this capability, we image a metallic
 371 plate carrying a dielectric “T” target through two scattering media: thin diffuser layers with
 372 varying grit levels (Thorlabs DG20-1500, DG10-600, DG20-220, and DG10-120) and thicker
 373 semi-translucent polymethyl methacrylate (PMMA) acrylic slabs with varying thicknesses (1,
 374 1.5, 2, and 3 mm). Scattering strength is quantified by transmittance, and additionally by optical
 375 thickness (OT) for PMMA slabs. Fig. 10 presents the weakest and strongest scattering endpoints

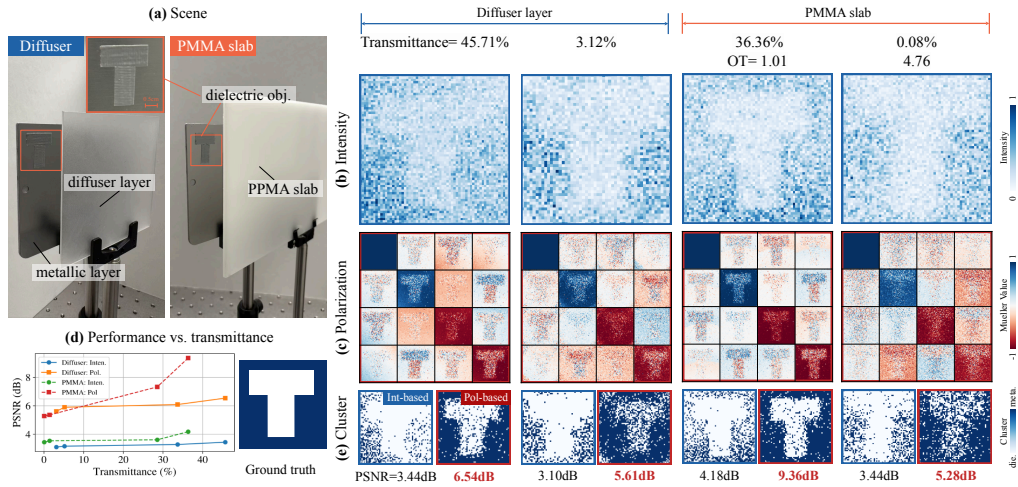


Fig. 10. Polarimetric imaging through scattering media with varying scattering strength. (a) A dielectric “T” target is attached to a metallic plate and imaged through either thin diffuser layers or semi-transparent PMMA acrylic slabs. (b) Intensity reconstructions under representative weak and strong scattering conditions: diffuser layers with transmittance of 45.71% and 3.12%, and PMMA slabs with transmittance of 36.36% and 0.08%, corresponding to OT of 1.01 and 4.76. (c) Reconstructed Mueller-matrix responses, which preserve material-dependent contrast between the metallic background and dielectric target. (d) Quantitative material-recovery performance versus transmittance, with PSNR computed against the binary “T” ground-truth mask. (e) Material clustering results using intensity-based and polarization-based features. Polarization-based clustering consistently improves dielectric–metallic separation, while performance degrades as scattering becomes stronger.

376 from each medium, with the complete sweep over all conditions provided in Supplement 1,
 377 Section 5.D. Because the experiment is performed in a reflective imaging-through-scattering
 378 geometry, the light traverses the scattering medium twice, making the reconstruction more
 379 challenging. Under these conditions, the intensity reconstructions provide weak material contrast
 380 because the dielectric target and metallic background have similar albedo and the scattering
 381 medium further blurs the coherent speckle (Fig. 10(b)). In contrast, the reconstructed Mueller
 382 responses retain clearer dielectric–metallic differences (Fig. 10(c)).

383 We evaluate material recovery by applying the same two-component clustering model [40]
 384 to either intensity features or Mueller-matrix features. The resulting segmentation is compared
 385 against a binary “T” mask obtained from a reference scan without the scattering medium. Across
 386 both diffuser and PMMA conditions, polarization-based clustering outperforms intensity-based
 387 clustering, while the advantage decreases as scattering becomes stronger (Fig. 10(d–e)).

388 6. Conclusion

389 Our work repurposes a compact telecommunications-grade optical modem for polarimetric full-
 390 wavefield coherent lidar. We introduce a polarimetric image formation model and reconstruction
 391 framework, and show that from a single coherent measurement per point we can recover not only
 392 depth and velocity but also an effective surface backscattering Jones response observed through
 393 the system’s illumination and collection aperture. By using our system to scan an entire scene, we
 394 can also perform spatial polarization analysis based on polarization speckle statistics, which goes
 395 beyond what conventional intensity-only or conventional coherent lidar systems can measure.
 396 Finally, our work shows the compatibility of telecommunications technology with coherent lidar.

397 Our investigation opens several directions for future work. Coherent speckle still limits the
398 fidelity of the recovered Jones matrices, and the scanning-based acquisition process remains
399 time-consuming (see Supplement 1, Section 5.F). Addressing these issues—e.g., by incorporating
400 speckle denoising, learning-based priors, and real-time processing—could improve reconstruction
401 fidelity and reduce latency. We hope that our work spurs further investigation into compact,
402 low-cost, polarization-resolved coherent lidar systems and algorithms that leverage off-the-shelf
403 photonic components. More broadly, advances in this area could enable new forms of lidar and
404 multimodal perception, improving sensing in low-visibility conditions for autonomous driving
405 and remote sensing.

406 **Acknowledgment.** We thank Ciena Corporation for loaning the optical modem. DBL and KNK
407 acknowledge support from Ciena Corporation, the Canada Foundation for Innovation, the Ontario Research
408 Fund, and NSERC under the RGPIN, and Alliance programs. DD was supported by the Faculty of Arts &
409 Science Postdoctoral Fellowship Award. PM was supported by the Mitacs Accelerate program. SB was
410 partly supported by the National Research Foundation of Korea (NRF) (No. RS-2024-00438532 and No.
411 RS-2023-00211658), the Institute of Information & Communications Technology Planning & Evaluation
412 (IITP) grant (No. RS-2024-0045788), and the IITP-ITRC (Information Technology Research Center) grant
413 (IITP-2026-RS-2024-00437866), all funded by the Korea government (MSIT).

414 **Disclosures.** The authors declare no conflicts of interest.

415 **Supplemental document.** See Supplement 1 for supporting content.

416 References

- 417 1. K. Kikuchi, “Fundamentals of coherent optical fiber communications,” *J. Light. Technol.* **34**, 157–179 (2015).
- 418 2. S. Gao, M. O’ Sullivan, and R. Hui, “Complex-optical-field LiDAR system for range and vector velocity measurement,”
419 *Opt. Express* **20**, 25867–25875 (2012).
- 420 3. S. W. Henderson, P. Gatt, D. Rees, and R. M. Huffaker, “Wind LiDAR,” in *Laser Remote Sensing*, (2005), pp.
421 487–740.
- 422 4. T. Wei, H. Xia, B. Yue, *et al.*, “Remote sensing of raindrop size distribution using the coherent Doppler LiDAR,” *Opt.*
423 *Express* **29**, 17246–17257 (2021).
- 424 5. Y. Li and J. Ibanez-Guzman, “LiDAR for autonomous driving: The principles, challenges, and trends for automotive
425 LiDAR and perception systems,” *IEEE Signal Process. Mag.* **37**, 50–61 (2020).
- 426 6. C. V. Poulton, M. J. Byrd, P. Russo, *et al.*, “Long-range LiDAR and free-space data communication with high-
427 performance optical phased arrays,” *IEEE J. Sel. Top. Quantum Electron.* **25**, 1–8 (2019).
- 428 7. D. H. Goldstein, *Polarized Light* (CRC Press, 2017).
- 429 8. J. W. Goodman, *Speckle Phenomena in Optics: Theory and Applications* (Roberts and Company Publishers, 2007).
- 430 9. J. R. Serres, P.-J. Lapray, S. Viollet, *et al.*, “Passive polarized vision for autonomous vehicles: A review,” *Sensors* **24**,
431 3312 (2024).
- 432 10. R. Blin, S. Ainouz, S. Canu, and F. Meriaudeau, “Road scenes analysis in adverse weather conditions by polarization-
433 encoded images and adapted deep learning,” in *Proc. IEEE Intell. Transp. Syst. Conf. (ITSC)*, (2019), pp. 27–32.
- 434 11. Z. Zhu, X. Li, J. Zhai, and H. Hu, “PODB: A learning-based polarimetric object detection benchmark for road scenes
435 in adverse weather conditions,” *Inf. Fusion* **108**, 102385 (2024).
- 436 12. C. Taglione, J. Guzzi, B. Cannelle, and D. Floreano, “Polarimetric imaging for robot perception: A review,” *Sensors*
437 **24**, 4440 (2024).
- 438 13. A. Kalra, V. Taamazyan, S. K. Rao, *et al.*, “Deep polarization cues for transparent object segmentation,” in *Proc.*
439 *IEEE/CVF Conf. Comput. Vis. Pattern Recognit.*, (2020), pp. 8602–8611.
- 440 14. J. Jiang, G. Cao, J. Deng, *et al.*, “Robotic perception of transparent objects: A review,” *IEEE Trans. Artif. Intell.* **5**,
441 2547–2567 (2024).
- 442 15. O. Morel and P. Gorria, “Polarization imaging for 3d inspection of highly reflective metallic objects,” *Opt. Spectrosc.*
443 **101**, 11–17 (2006).
- 444 16. F. Meriaudeau, M. Ferraton, C. Stolz, *et al.*, “Polarization imaging for industrial inspection,” in *Proc. SPIE Image*
445 *Process. Mach. Vis. Appl.*, vol. 6813 (SPIE, 2008), p. 681308.
- 446 17. Q. Liu, S. Wu, B. Liu, *et al.*, “Shipborne variable-fov, dual-wavelength, polarized ocean LiDAR: Design and
447 measurements in the Western Pacific,” *Opt. Express* **30**, 8927–8948 (2022).
- 448 18. A. P. Vasilkov, Y. A. Goldin, B. A. Gureev, *et al.*, “Airborne polarized LiDAR detection of scattering layers in the
449 ocean,” *Appl. Opt.* **40**, 4353–4364 (2001).
- 450 19. X. Liu, L. Zhang, X. Zhai, *et al.*, “Polarization LiDAR: Principles and applications,” *Photonics* **10**, 1118 (2023).
- 451 20. M. Amani, A. Moghimi, S. M. Mirmazloumi, *et al.*, “Ocean remote sensing techniques and applications: A review
452 (part i),” *Water* **14**, 3400 (2022).
- 453 21. J. Zheng, *Optical Frequency-Modulated Continuous-Wave (FMCW) Interferometry* (Springer, 2005).

- 454 22. S. Gao and R. Hui, "Frequency-modulated continuous-wave LiDAR using I/Q modulator for simplified heterodyne
455 detection," *Opt. Lett.* **37**, 2022–2024 (2012).
- 456 23. L. Zhang, J. Sun, H. Cong, *et al.*, "Polarimetric imaging LiDAR based on FMCW and Mueller matrix for target
457 classification," *J. Light. Technol.* (2024).
- 458 24. N. Takeuchi, H. Baba, K. Sakurai, and T. Ueno, "Diode-laser random-modulation CW LiDAR," *Appl. Opt.* **25**, 63–67
459 (1986).
- 460 25. C. S. Sambridge, J. T. Spollard, A. J. Sutton, *et al.*, "Detection statistics for coherent RMCW LiDAR," *Opt. Express*
461 **29**, 25945–25959 (2021).
- 462 26. A. Lukashchuk, J. Riemensberger, A. Tusnin, *et al.*, "Chaotic microcomb-based parallel ranging," *Nat. Photonics* **17**,
463 814–821 (2023).
- 464 27. P. Tang, M. A. Kirby, N. Le, *et al.*, "Polarization-sensitive optical coherence tomography with single input for imaging
465 depth-resolved collagen organizations," *Light Sci. Appl.* **10**, 237 (2021).
- 466 28. J. F. De Boer, C. K. Hitzenberger, and Y. Yasuno, "Polarization-sensitive optical coherence tomography—a review,"
467 *Biomed. Opt. Express* **8**, 1838–1873 (2017).
- 468 29. Y. Sun, J. Wang, J. Shi, and S. A. Boppart, "Synthetic polarization-sensitive optical coherence tomography by deep
469 learning," *NPJ Digit. Med.* **4**, 105 (2021).
- 470 30. P. Mirdehghan, B. Buscaino, M. Wu, *et al.*, "Coherent optical modems for full-wavefield LiDAR," in *Proc. SIGGRAPH*
471 *Asia*, (2024).
- 472 31. E. Ip, A. P. T. Lau, D. J. F. Barros, and J. M. Kahn, "Coherent detection in optical fiber systems," *Opt. Express* **16**,
473 753–791 (2008).
- 474 32. B. E. A. Saleh and M. C. Teich, *Fundamentals of Photonics, 2-Volume Set* (John Wiley & Sons, 2019).
- 475 33. A. Yariv and P. Yeh, *Photonics: Optical Electronics in Modern Communications*, vol. 6 (Oxford University Press,
476 2007).
- 477 34. G. P. Agrawal, *Fiber-Optic Communication Systems* (John Wiley & Sons, 2012).
- 478 35. R. Chipman, W. S. T. Lam, and G. Young, *Polarized Light and Optical Systems* (CRC Press, 2018).
- 479 36. M. Takeda, W. Wang, and S. G. Hanson, "Polarization speckles and generalized stokes vector wave: a review," in
480 *Proc. SPIE Speckle: Int. Conf. Speckle Metrol.*, vol. 7387 (2010), pp. 257–263.
- 481 37. R. C. Jones, "A new calculus for the treatment of optical systems. i. description and discussion of the calculus," *J.*
482 *Opt. Soc. Am.* **31**, 488–493 (1941).
- 483 38. S. T. Tang and H. S. Kwok, " 3×3 matrix for unitary optical systems," *J. Opt. Soc. Am. A* **18**, 2138–2145 (2001).
- 484 39. S.-H. Baek and F. Heide, "All-photon polarimetric time-of-flight imaging," in *Proc. IEEE/CVF Conf. Comput. Vis.*
485 *Pattern Recognit.*, (2022), pp. 17876–17885.
- 486 40. C. M. Bishop, *Pattern Recognition and Machine Learning* (Springer, 2006).